

Performance and Analysis of Sensor less BLDC Motor Drive with Fuzzy Controller

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Abstract: In this paper the performance of PID/fuzzy controller-based sensing element less BLDC servomotor drives is investigated beneath totally different operative conditions like modification in reference speed, parameter variations, load disturbance, etc. the tactic of Back electromotive force sensing is employed for providing the controlled gate pulse to the electrical converter. This drive System overcomes the restrictions of Hall Sensors exploitation back electromotive force management exploitation Fuzzy / inflammatory disease controller-based sensing element less BLDC servomotor drives. Brushless motors square measure normally used as pump, fan and spindle drives in adjustable or variable speed applications.

Keywords— Servomotor drive, fuzzy controller, modelling, Sensor less Brushless dc (BLDC), transient and steady-state performance.

I. INTRODUCTION

The brushless DC (BLDC) engine is turning out to be progressively famous in divisions, for example, car especially electric vehicles , HVAC, white products and modern on the grounds that it gets rid of the mechanical commutator utilized as a part of conventional engines, supplanting it with an electronic gadget that enhances the unwavering quality and solidness of the unit. The brushes of a traditional engine transmit energy to the rotor windings which, when stimulated, turn in a settled attractive field. Erosion between the stationary brushes and a turning metal contact on the turning rotor causes wear. What's more, power can be lost because of poor brush to metal contact and arcing. BLDC engines, additionally called Permanent Magnet DC Synchronous engines, are one of the engine sorts that have all the more quickly picked up prominence, predominantly due to their better attributes and execution. These engines are utilized as a part of an awesome measure of mechanical divisions in light of the fact that their structural engineering is suitable for any security basic applications. The brushless DC engine is a synchronous electric engine from a displaying point of view, looks precisely like a DC engine, having a direct relationship in the middle of current and torque, voltage and speed.

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It is an electronically controlled recompense framework, rather than having a mechanical compensation, which is commonplace of brushed engines. Moreover, the electromagnets don't move, the lasting magnets turn and the armature stays static. This gets around the issue of how to exchange current to a moving armature. Keeping in mind the end goal to do this, the brush-framework/commutator get together is supplanted by a canny electronic controller, which performs the same force appropriation as a brushed DC engine. BLDC engines have numerous gets each point of preference of a DC machine focal points over brushed DC engines and actuation engines, for example, a superior rate versus torque attributes, high element reaction, high proficiency and unwavering quality, long working life (no brush erosion), noiseless operation, higher rate ranges, and decrease of electromagnetic obstruction EMI. Brushless dc engine is one sort of perpetual magnet synchronous engine, having changeless magnets on the rotor and trapezoidal shape back EMF. The BLDC engine utilizes a dc control supply changed to the stator stage windings of the engine by force gadgets, the exchanging arrangement being resolved from the rotor position.

The stage control of BLDC engine, fit as a fiddle, is synchronized with the back EMF to create steady torque at a consistent rate.

The mechanical commutator of the brush dc engine is supplanted by electronic switches, which supply current to the engine windings as a component of the rotor position. These brushless dc engines are by and large controlled utilizing a three stage inverter, requiring a rotor position sensor for beginning and for giving the best possible compensation succession to control the inverter. These position sensors can be lobby sensors, or total position sensors. An ordinary BLDC engine control framework with position sensors. Those sensors will build the expense and size of the engine, and an exceptional mechanical course of action should be made for mounting sensors.

The control of BLDC engines should be possible in sensor or sensor less mode, however to decrease general expense of inciting gadgets, sensor less control methods are ordinarily utilized. The upside of sensor less BLDC engine control is that the identifying part can be avoided, and in this manner over all expenses can be extensively lessened. The disservices of sensor less control are higher prerequisites for control calculations and more confused hardware. The majority of the electrical engines that don't require an electrical connection (made with brushes) in the middle of stationary and pivoting parts can be considered as brushless changeless

magnet (PM) machines, which can be classified in view of the PMs mounting and the back-EMF shape.

II. MODELLING OF BLDC SERVOMOTOR DRIVE

The BLDC servomotor drive framework comprising of BLDC servomotor and IGBT inverter is demonstrated in view of the suppositions that all stator stage windings have approach resistance per stage, steady self and common inductances; power semiconductor switches are perfect; iron misfortunes are insignificant; and the engine is unsaturated. The proportionate circuit of the BLDC servomotor drive framework is appeared.

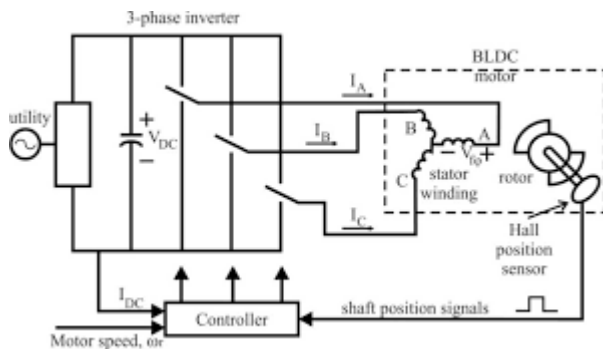


Fig. 1. Modeling of BLDC servomotor

The line to line voltage mathematical statements are communicated in matrix form as $V = X +$ (1)

Since basic inductance is irrelevant when appeared differently in relation to the self-inductance, the above correlation can be revised as $V = X +$ (2)

Where L and M are self-inductance and common inductance per stage; R is the stator winding resistance per stage; e_a , e_b and e_c are the back EMFs of stages a , b and c separately; i_a , i_b , and i_c are the stage streams of stage a , b and c individually. The electromagnetic torque created by the engine can be communicated as

$$T_e = (e_a i_a + e_b i_b + e_c i_c) / \omega = K_t I \quad (3)$$

Where $i_a = i_b = i_c = I$, ω is the rakish speed in radians every second, and K_t is the torque consistent. Since the electromagnetic torque is used to defeat the restricting Torques of latency and load, it can likewise be composed as

$$T_e = T_L + J_M + B_M \omega \quad (4)$$

Where T_L is the heap torque, J_M is the dormancy, and B_M is the rubbing consistent of the BLDC servomotor. The load torque can be communicated as far as burden idleness J_L and grinding B_L parts as

$$T_L = J_L + B_L \omega \quad (5)$$

The yield power created by the engine is

$$P = T_e \omega \quad (6)$$

$$E = e_a = e_b = e_c = K_b \omega \quad (7)$$

Where K_b is back EMF steady, E is back EMF per stage, and ω is precise speed in radians every second. The parameters of engine are stage resistance, stage inductance, and idleness and contact of BLDC servomotor and load. It is important to decide the parameters of both BLDC servomotor and stack in order to outline traditional controllers like P, PI, and PID controllers.

These parameters can impact the pace reaction of the BLDC servomotor drive framework. Increment in the estimation of vitality stockpiling inactivity components J_M and J_L will increment settling time of the velocity reaction or the other way around. The decline in the estimations of force expending grating parts B_m and B_L will build the deceleration time of the rate reaction.

III. SENSORLESS BLDC MOTOR

A way to deal with position sensorless BLDC engine drive, another calculation for sensorless operation and sensorless control without sign infusion are accounted for. Two sorts of sensorless control systems of Permanent magnet BLDC engines are talked about in.

- Back EMF Detection strategy
- Position estimation utilizing engine parameters

The position estimation plot more often than not needs muddled calculation, and the expense of the framework is moderately high. The back EMF detecting plan is suitable to be utilized on an extensive variety.

A. Back EMF Detection method

The corridor sensor signal changes the state when the voltage extremity of back EMF crosses from positive to negative or negative to positive. In perfect cases, this happens on zero-intersection of back EMF, yet basically, there will be a deferral because of the winding qualities. This postponement ought to be remunerated by the controller. Since back EMF is relative to the velocity of turn, at a low speed, the back EMF would be at low plentifulness to identify zero-intersection.

The accompanying suppositions are considered for sensorless BLDC engine. The engine is not soaked. Stator resistances of the considerable number of windings are equivalent R_S , self inductances are steady L_S and shared inductances M are zero. Iron misfortunes are unimportant.

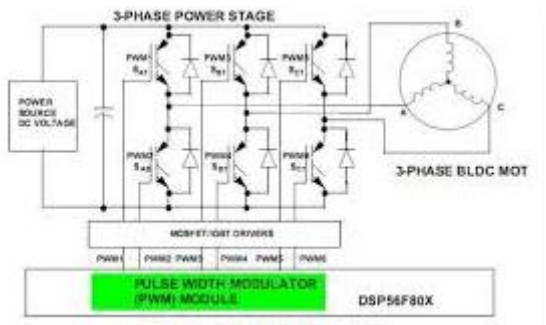


Fig. 2. Sensorless BLDC engine drive

The engine must be begun in open circle, from stop and when adequate back EMF is assembled to distinguish the zero-cross point, the control ought to be moved to the back EMF detecting. The base velocity at which back EMF can be detected is computed from the back EMF steady of the engine. With this technique for replacement, the corridor sensors can be wiped out and in a few engines, the magnets for lobby sensors additionally can be disposed of.

B. Back EMF zero intersection estimation system

The stage A terminal voltage regarding the star purpose of the stator Van is given as

$$V_{an} = R_a i_a + L_a \frac{di_a}{dt} + e_{an} \tag{8}$$

Where Ra is the stator resistance, La is the stage inductance; ean is the back EMF, and ia is the stage current of stage A.

Comparable mathematical statements can be composed for the other two stages, are

$$V_{bn} = R_b i_b + L_b \frac{di_b}{dt} + e_{bn} \tag{9}$$

$$V_{cn} = R_c i_c + L_c \frac{di_c}{dt} + e_{cn} \tag{10}$$

Where the images have their undeniable implications. From this, line voltage Vab may be resolved as

$$V_{ab} = V_{an} - V_{bn} = R(i_a - i_b) + L_d \frac{d(i_a - i_b)}{dt} + e_{an} - e_{bn} \tag{11}$$

Correspondingly

$$V_{bc} = R(i_b - i_c) + L_d \frac{d(i_b - i_c)}{dt} + e_{bn} - e_{cn} \tag{12}$$

$$V_{ca} = R(i_c - i_a) + L_d \frac{d(i_c - i_a)}{dt} + e_{cn} - e_{an} \tag{13}$$

$$V_{abbc} = R(i_a - 2i_b + i_c) + L_d \frac{d(i_a - 2i_b + i_c)}{dt} + e_{an} - 2e_{bn} + e_{cn} \tag{14}$$

Considered the interim stage An and C are directing and stage B is open as showed by the shaded area .

$$V_{abbc} = V_{ab} - V_{bc} = e_{an} - 2e_{bn} + e_{cn} = - 2e_{bn} \tag{15}$$

The distinction of line voltages waveform is a modified representation of the back EMF waveform. The back-EMF of skimming stage is detected and its zero intersection is recognized by contrasting it and nonpartisan point voltage. This plan experiences high normal mode voltage and high recurrence commotion because of the PWM drive, so it requires low pass channels, and voltage dividers.

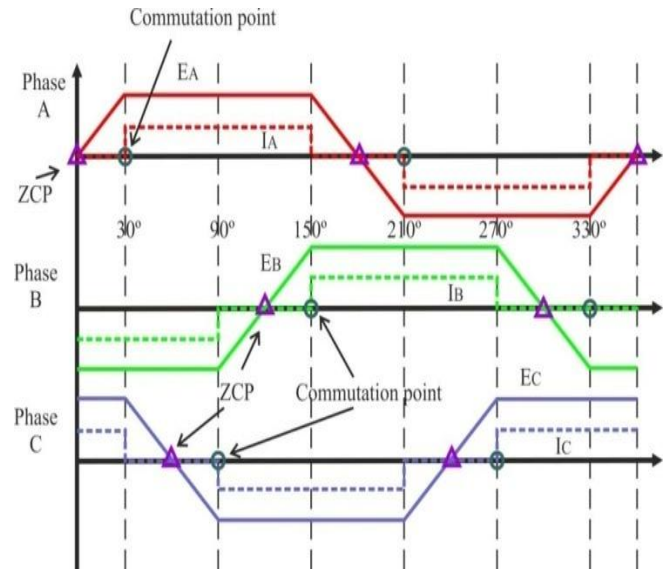


Fig. 3. Back EMF waveforms of phase A, B and C

IV. DESIGN AND IMPLEMENTATION OF FUZZY CONTROLLER

Fuzzy logic enables to describe the general behaviour of the system in a linguistic manner by forming IF-THEN rules that are in the form of statements. The great challenge is to design and implement the FLC quickly by framing minimum number of rules based on the knowledge of the system. The block diagram of fuzzy inference operation is shown below.

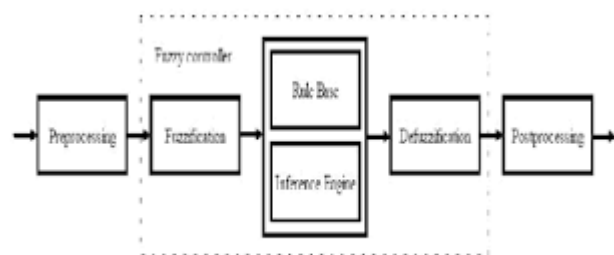


Fig. 4. Block diagram of fuzzy inference system

The general FLC consists of four parts they are fuzzification, fuzzy rule base, fuzzy inference engine, and defuzzification. The design steps are as follows.

A. Defining fuzzy membership functions and rules

To perform fuzzy computation, the inputs must be converted from numerical or crisp value into fuzzy values and the output should be converted from fuzzy value to crisp. The fuzzy input variables “Error” and “Change in error” are quantized using the following linguistic terms negative N, Zero Z, and positive P. the fuzzy output variable “Change in Duty cycle” is quantized using the following linguistic terms decrease D, No change NC, and increase I. fuzzy membership functions are used as tools to convert crisp values to linguistic terms.

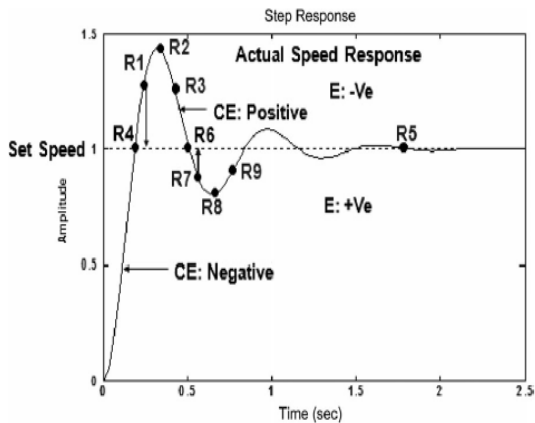


Fig. 5. Illustration for the formation of rules for a typical underdamped BLDC servomotor drive system.

| E \ CE | NME | NSE | NLE | ZE | PME | PSE | PLE |
|--------|-----|-----|-----|----|-----|-----|-----|
| NMCE | D | D | D | D | 1 | 1 | 1 |
| NSCE | D | D | D | D | 1 | 1 | 1 |
| NLCE | D | D | D | D | 1 | 1 | 1 |
| ZCE | D | D | D | NC | 1 | 1 | 1 |
| PMCE | D | D | D | 1 | 1 | 1 | 1 |
| PSCE | D | D | D | 1 | 1 | 1 | 1 |
| PLCE | D | D | D | 1 | 1 | 1 | 1 |

Table No.1

R1: If Error E is Negative NME and change in error CE is Negative NMCE THEN change in duty cycle ΔDC is decrease D. This rule implies that when the system output is at R1, then the actual speed is greater than the reference speed and the motor is accelerating, so the duty cycle of the IGBTs of the inverter module should be decreased to reduce the

average voltage applied across the phase windings and bring the actual speed of the system close to the reference speed.

R2: If Error E is Negative NME and change in error CE is Negative NSCE THEN change in duty cycle ΔDC is decrease D.

R3: If Error E is Negative NME and change in error NLCE is Negative THEN change in duty cycle ΔDC is decrease D.

R4: If Error E is Negative NME and change in error ZCE is Zero THEN change in duty cycle ΔDC is decrease D. This rule implies that when the system output is at R5, then there should be a no change in the duty cycle as the actual speed has already reached the steady state.

R5: If Error E is Negative NME and change in error CE is Positive PMCE THEN change in duty cycle ΔDC is decrease D.

R6: If Error E is Negative NME and change in error CE is Positive PSCE THEN change in duty cycle ΔDC is decrease D.

R7: If Error E is Negative NME and change in error CE is Positive PLCE THEN change in duty cycle ΔDC is decrease D. This rule implies that when the system output is at R7, then the actual speed is lesser than the reference speed and the motor is decelerating, so the duty cycle of the IGBTs of the inverter module should be increased so as to increase the average voltage applied across the phase windings and bring the actual speed of the system close to reference speed.

R8: If Error E is Negative NSE and change in error CE is Negative NMCE THEN change in duty cycle ΔDC is decrease D.

R9: If Error E is Negative NSE and change in error CE is Negative NSCE THEN change in duty cycle ΔDC is decrease D.

R10: If Error E is Negative NSE and change in error CE is Negative NLCE THEN change in duty cycle ΔDC is decrease D.

R11: If Error E is Negative NSE and change in error CE is Zero ZCE THEN change in duty cycle ΔDC is decrease D.

R12: If Error E is Negative NSE and change in error CE is Positive PMCE THEN change in duty cycle ΔDC is decrease D.

R13: If Error E is Negative NSE and change in error CE is Positive PSCE THEN change in duty cycle ΔDC is decrease D.

R14: If Error E is Negative NSE and change in error CE is Positive PLCE THEN change in duty cycle ΔDC is decrease D.

R15: If Error E is Negative NLE and change in error CE is Negative NMCE THEN change in duty cycle ΔDC is decreased D.

- R16: If Error E is Negative NLE and change in error CE is Negative NSCE THEN change in duty cycle DC is decrease D.
- R 17: If Error E is Negative NLE and change in error CE is Negative NLCE THEN change in duty cycle ΔDC is decrease D.
- R 18: If Error E is Negative NLE and change in error CE is Zero ZCE THEN change in duty cycle ΔDC is decrease D.
- R 19: If Error E is Negative NLE and change in error CE is Positive PMCE THEN change in duty cycle ΔDC is decrease D.
- R 20: If Error E is Negative NLE and change in error CE is Positive PMCE THEN change in duty cycle ΔDC is decrease D.
- R 21: If Error E is Negative NLE and change in error CE is Positive PLCE THEN change in duty cycle ΔDC is decrease D.
- R 22: If Error E is Zero ZE and change in error CE is Negative NMCE THEN change in duty cycle ΔDC is decrease D.
- R 23: If Error E is Zero ZE and change in error CE is Zero NSCE THEN change in duty cycle ΔDC is decrease D.
- R 24: If Error E is Zero ZE and change in error CE is Zero NLCE THEN change in duty cycle ΔDC is decrease D.
- R 25: If Error E is Zero ZE and change in error CE is Zero ZCE THEN change in duty cycle ΔDC is No change NC.
- R 26: If Error E is Zero ZE and change in error CE is Positive PMCE THEN change in duty cycle ΔDC is Increased I.
- R 27: If Error E is Zero ZE and change in error CE is Positive PSCE THEN change in duty cycle ΔDC is Increased I.
- R 28: If Error E is Zero ZE and change in error CE is Positive PLCE THEN change in duty cycle ΔDC is Increased I.
- R 29: If Error E is Positive PME and change in error CE is Negative NMCE THEN change in duty cycle ΔDC is increase I.
- R 30: If Error E is Positive PME and change in error CE is Negative NSCE THEN change in duty cycle ΔDC is increase.

V. SIMULATION OF SENSORLESS BRUSHLESS DC MOTOR DRIVE

The simulation model of Sensorless Brushless DC Motor drive was created by the mathematical models. This model was simulated with PID and Fuzzy controllers. Back emf detection method of Sensorless BLDC drive was controlled by controllers.

The Back emf of BLDC motor is sensed using voltage sensing block and it generates a gate pulse to switch on the three phase inverter for different back emf generated in the motor. The PID and FUZZY Controllers sense the disturbance occur in BLDC motor and made the motor to run without any change in parameters due to the load disturbance

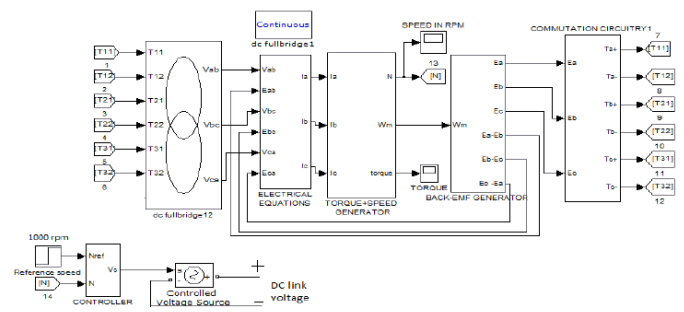
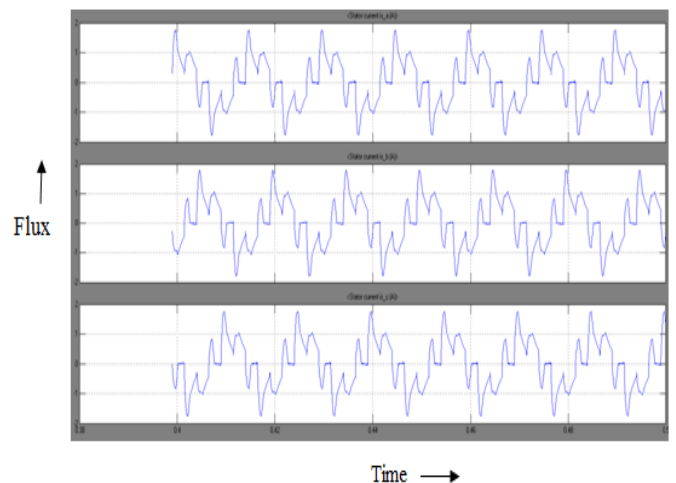


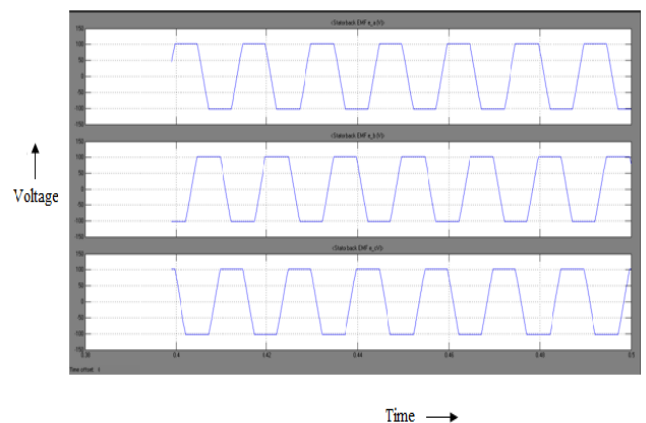
Fig. 6. Simulation circuit of Sensorless BLDC motor drive

A. Simulation result for Stator current

It is found that the peak amplitude of phase back EMF increases when there is a sudden decrease in load. This is because the speed increases due to sudden decrease in load. Similarly, the peak amplitude of phase back EMF decreases, when the speed decreases because of a sudden increase in load.

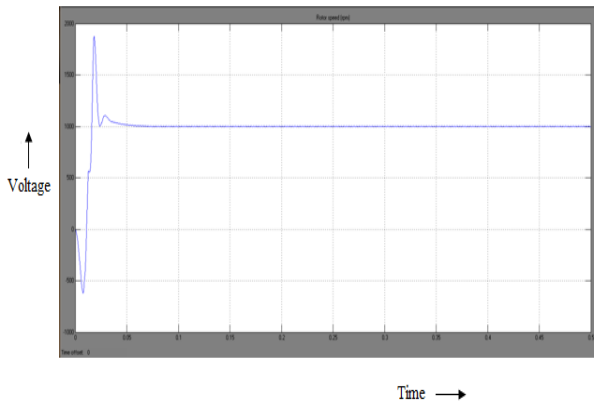


B. Simulation result for back emf generated in the BLDC motor drive



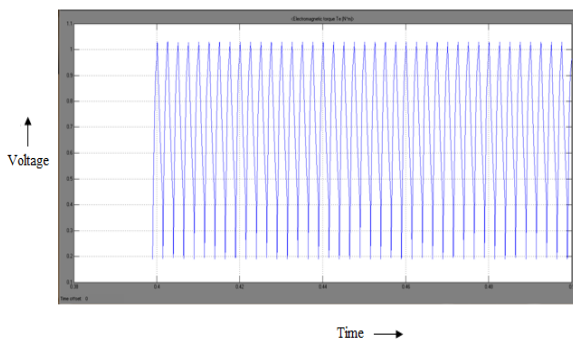
Speed response of the Fuzzy controller-based BLDC servomotor drive for step change in reference speed with system parameters $J1$, $R1$ and 100% Load. It is found that speed response always follows the change in reference speed and settles close to the reference speed.

C. Simulation result of rotor speed of BLDC motor drive



The speed response of fuzzy controller-based BLDC servomotor drive is found to be much faster than PID controller-based BLDC servomotor drive under load conditions.

D. Simulation Result of torque of BLDC motor drive



VI. CONCLUSION

The PID and fluffly control methods are effectively executed for sensorless BLDC servo engine drive framework. The heap reaction of fluffly controller based sensorless BLDC servomotor drive is like burden reaction of PID controller based sensorless BLDC servomotor drive when framework parameters are steady. Be that as it may, when framework parameters are not consistent the PID controller reaction are changed. Hence, PID controller based sensorless BLDC servomotor drive neglected to give enhanced execution under execution varieties of the framework. Yet, test comes about plainly demonstrates that fluffly controller based sensorless BLDC servomotor drive can give an enhanced pace reaction with reliably same ascent time, and settling time when the framework subjected to load aggravation, parameter varieties, and step change in reference speed.

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